

Cct Symbol of VFOA

I/p 1 - Non Inverting i/p

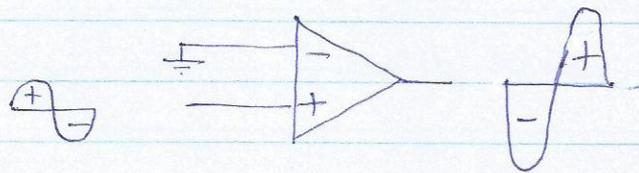
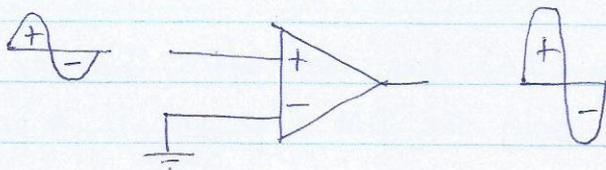
I/p 2 - Inverting i/p

Common VFOA Type

OA can be powered by → Dual DC supply ( $+V_s, -V_s$ ) - Dual Supply OAs

→ Single DC supply ( $V_s$ ) - Single Supply OAs.

Some are compatible with logic ccts.



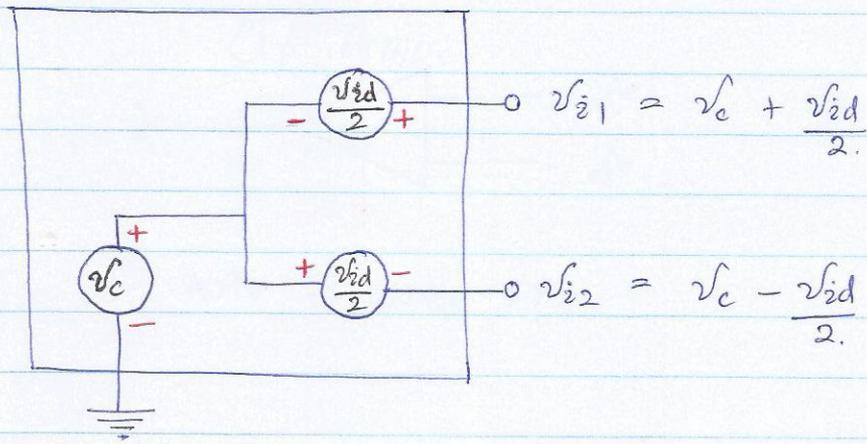
With OAs two types of i/p signals are associated with,

\* Differential Signals,  $v_{id}$  (Differential mode signals)

\* Common Signals,  $v_{ic}$  (Common mode signals)

I/Ps 1 and 2 are above the common (gnd) potential,

They are represented by the following model.



$$v_{i1} - v_{i2} = v_{id}$$

$$\frac{v_{i1} + v_{i2}}{2} = v_c$$

Usually,

- Common signal,  $v_c$  is unwanted
- Diff. mode signal is the wanted signal.

A Differential Amp (DA) has two i/p's. Similar to OA. An Ideal DA amplifies only the DM i/p signal. A practical DA amplifies more than DM signal and less the CM signal.

Two types of Gains can be defined:

$$\text{Differential Gain, (DM Gain)} \quad A_d = \frac{v_o}{v_{id}} = \frac{v_o}{v_{i1} - v_{i2}}$$

$$\text{Common mode Gain,} \quad A_c = \frac{v_o}{v_c} = \frac{v_o}{\left(\frac{v_{i1} + v_{i2}}{2}\right)}$$

$$\text{Common Rejection Ratio (CMRR)} = \left| \frac{A_d}{A_c} \right|$$

CMRR is specified in dB

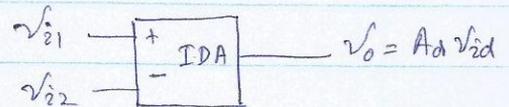
$$\text{CMRR}_{\text{dB}} = 20 \log_{10} \left| \frac{A_d}{A_c} \right|$$

For an IDA

$A_d$  - Finite

$A_c = 0$

$$\therefore \text{CMRR} = \infty$$

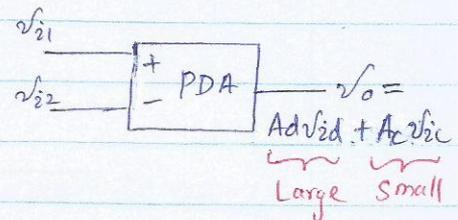


For a Practical DA

$A_d$  - Finite (High)

$A_c \neq 0$  (Low)

$$\therefore \text{CMRR} \neq \infty \text{ (high)}$$



Ex!  $A_d = 1000$

$\text{CMRR} = 80 \text{ dB}$

$\therefore A_c = ?$

$$\left| \frac{A_d}{A_c} \right| = 10^4 \Rightarrow A_c = \frac{A_d}{10^4} = \frac{1}{10} = \underline{\underline{0.1}}$$

OA also has,

high DM Gain,  $A_d$

low CM Gain,  $A_c$

$$v_o = A_d v_{id} + A_c v_{ic} = A_d \left[ v_{id} + \frac{v_{ic}}{\text{CMRR}} \right]$$

For general purpose IC OA (Ex: 741, TL082, ...)  $A_d$  is of the order of  $10^5$  (100 dB)

CMRR can be  $> 80 \text{ dB}$

$$\therefore v_o \approx A_d v_{id}$$

- OAs are manufactured using different technologies,

Ex: BJTs - Bipolar OAs (Ex: 741)

BIFET - BiFET OAs (Ex. TL082)

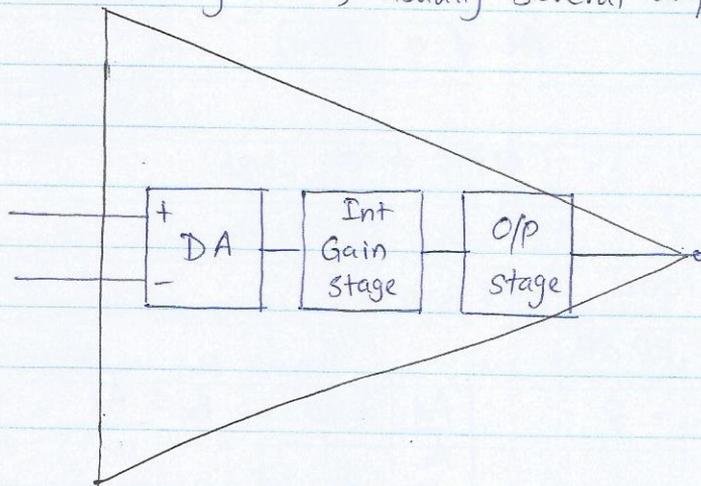
(FET+BJT)

CMOS - CMOS OAs

⋮ ⋮

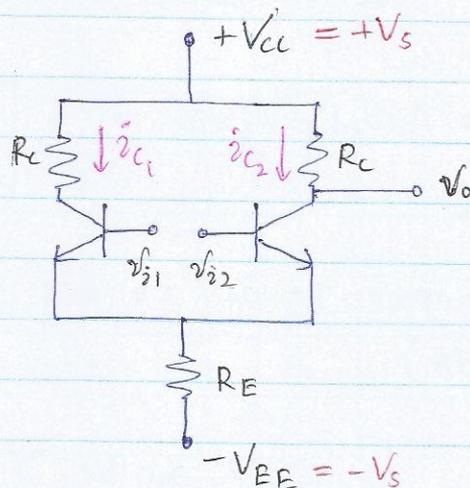
### Internal Architecture of OA

To achieve high  $A_d$ , usually several amplifying stages are used.

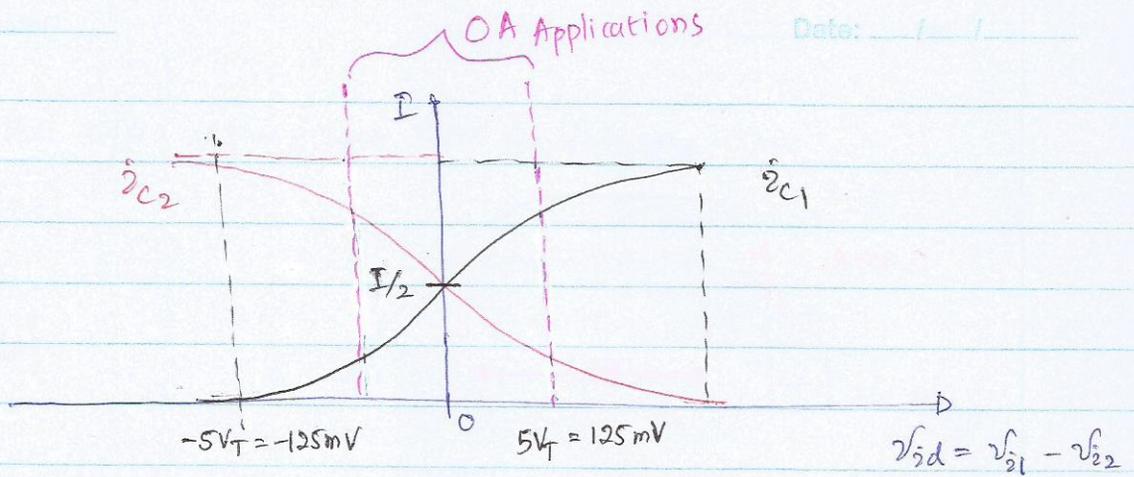


### Basic BJT DA

(Differential pair  
or  
Emitter Coupled Pair)



→ Basis for high speed logic family ECL



OA can be used,

- Open-Loop Configuration

Limited Applications

- Closed-Loop Configuration (with feedback)

Many Applications

Important Electrical Parameters of a general purpose IC OA

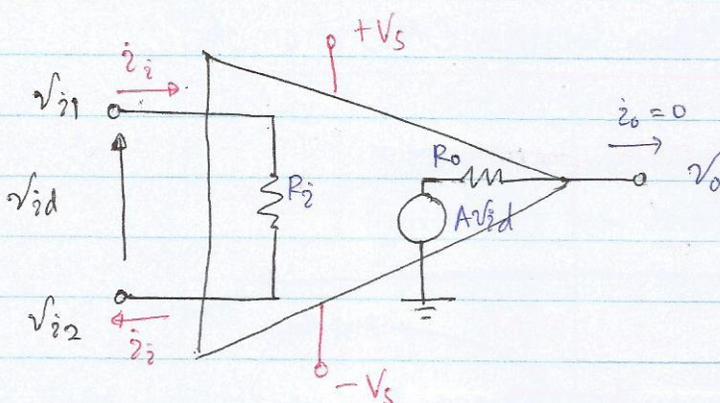
Basic parameters

Open loop gain,  $A = A_d = \frac{v_o}{v_{id}} = \frac{v_o}{v_{i1} - v_{i2}} = 100 \text{ dB} \approx 10^5 (2 \times 10^5)$

I/P Impedance,  $R_i = 1 \text{ M}\Omega (2 \text{ M}\Omega)$

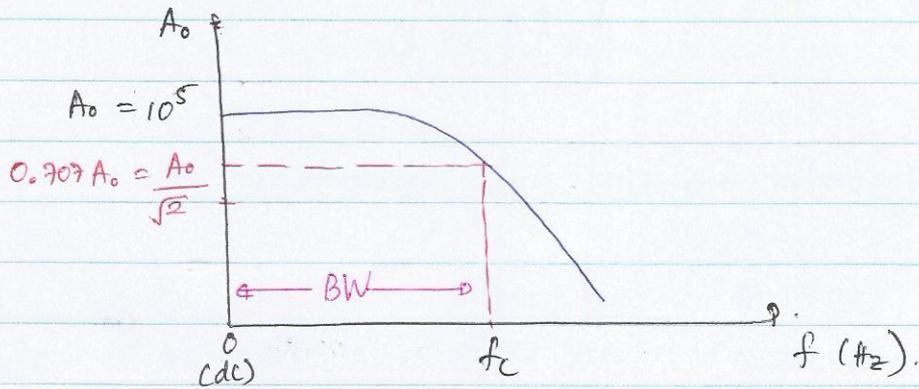
O/P Impedance,  $R_o = 100 \Omega ( )$

Bandwidth, BW =  $1 \text{ MHz} (1.5 \text{ MHz})$



$$R_i = \frac{v_{id}}{i_i} = \frac{v_{i1} - v_{i2}}{i_i}$$

$$R_o = \frac{v_o |_{oc}}{i_o |_{sc}}$$



### Ideal OA

A hypothetical concept, but important to simplify ckt. analysis.

For an IOA,

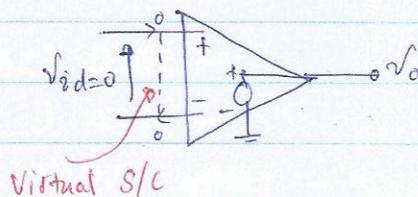
$$A = \infty \quad (\because \text{CMRR} = \infty) \implies v_{id} = 0$$

$$v_{i1} = v_{i2}$$

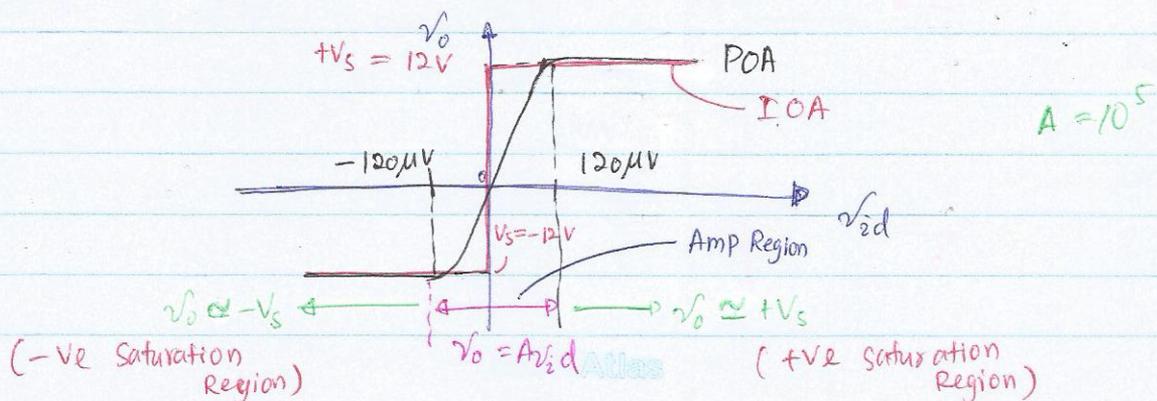
$$R_i = \infty \quad (i_i = 0)$$

$$R_o = 0$$

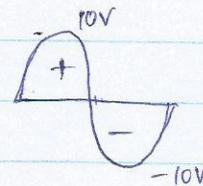
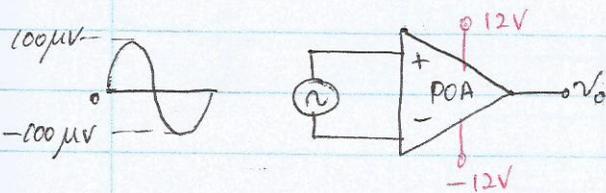
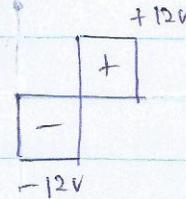
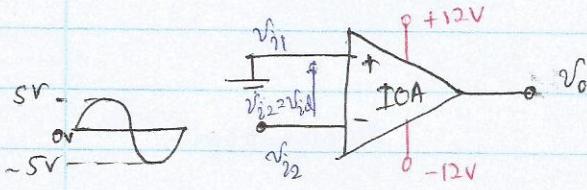
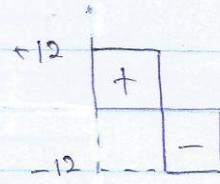
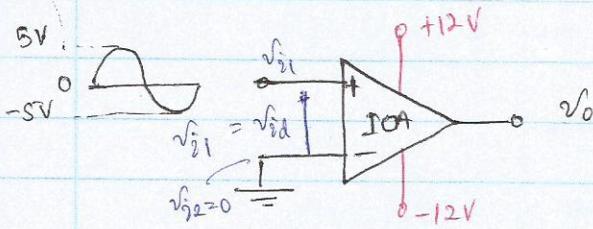
$$\text{BW} = \infty$$



### Voltage Transfer Characteristics (VTC) of an OA



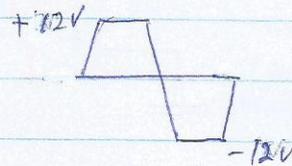
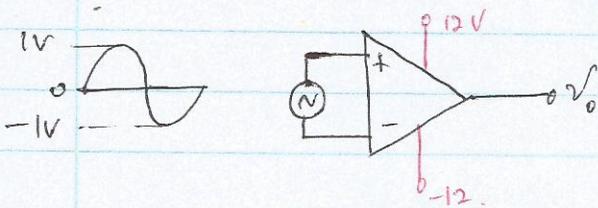
Q1 Sketch the o/p voltage wave form for following ccts.



$$V_o = A v_{id}$$

$$= 10^5 \times 100 \times 10^{-6}$$

$$= 10$$



Practical OP Amp ccts.

OP Amp based Amplifiers.

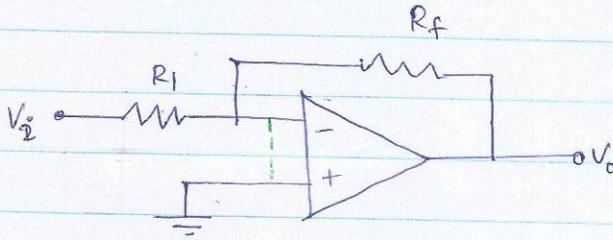
- Inverting Amplifiers (-ve gain)
- Non inverting Amplifiers. (+ve gain)
- Summing Amplifiers (OP Amp adder or Analogue Adder)
- Buffer Amplifiers (unity +ve gain amp)

Unless otherwise mentioned we may assume an IOA.

- In these applications OA is used in closed-loop Configuration. (Negative Feedback)

Part of the o/p is fed back to the i/p and added (-vely).

Inverting Amp -



$$\frac{v_i}{R_i} + \frac{V_o}{R_f} = 0$$

∴ Closed loop.

∴ Open loop

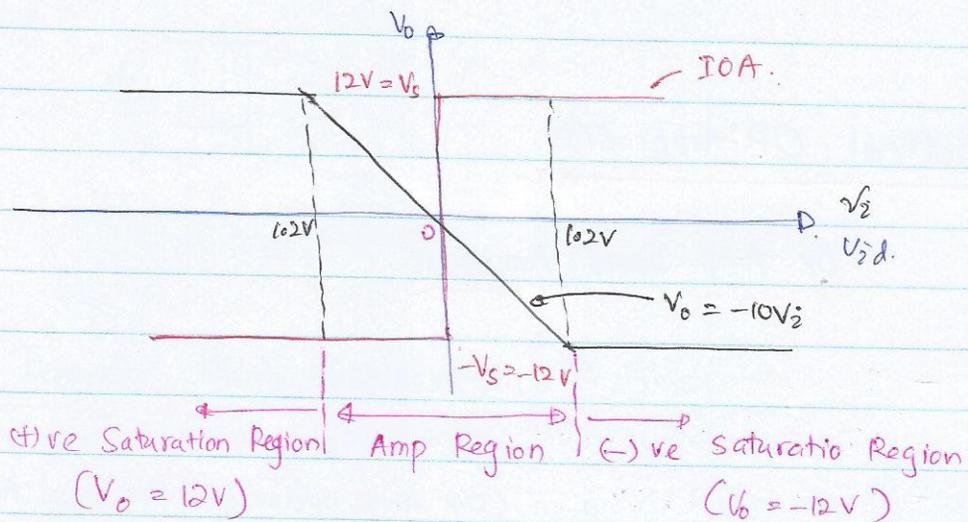
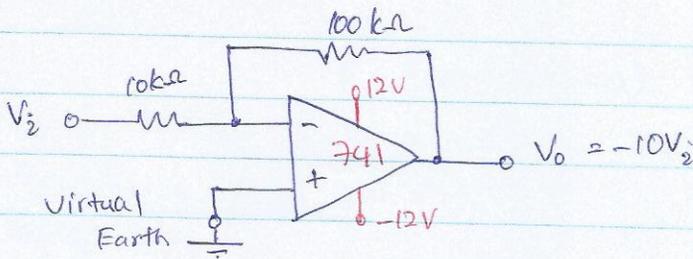
Voltage gain,

$$\frac{V_o}{V_i} = \frac{-R_f}{R_i}$$

Voltage gain,

$$\frac{V_o}{V_{id}} = \infty$$

ex.



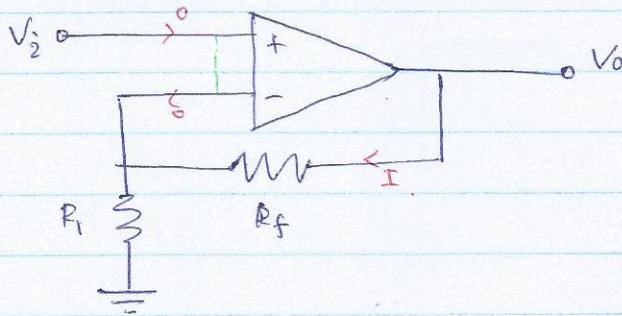
Q2 Sketch the output voltage wave for,

(i)  $V_i(t) = 0.5 \sin \omega t$  Volts —  $V_p = 0.5 < 1.2V$   
 (o/p not clipped)

$V_o(t) = -5 \sin \omega t$   
 $= 5 \sin(\omega t + 180^\circ)$

(ii)  $V_i(t) = 2 \sin \omega t$  Volts —  $V_p = 2V > 1.2V$   
 (o/p is clipped)

Non Inverting Amp



$$I = \frac{V_o - V_i}{R_f} = \frac{V_i}{R_1}$$

$\therefore$  Close Loop Gain,

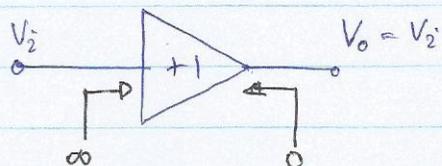
$$\frac{V_o}{V_i} = 1 + \frac{R_f}{R_1} \geq 1$$

Buffer Amp has,

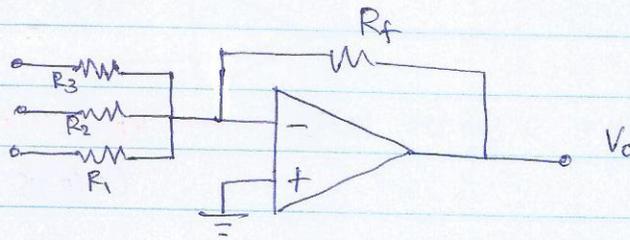
$$A_v = \frac{V_o}{V_i} = 1$$

$$R_i = \infty$$

$$R_o = 0$$



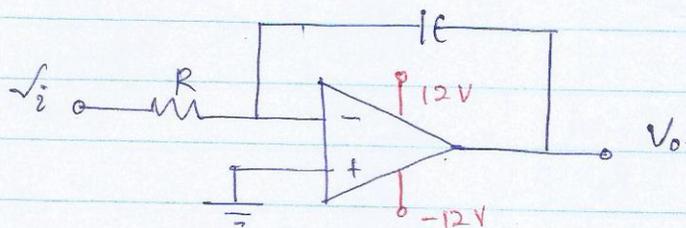
## Summing Amplifier



$$V_0 = - \left( \frac{R_f}{R_1} v_{i1} + \frac{R_f}{R_2} v_{i2} + \frac{R_f}{R_3} v_{i3} \right)$$

$$V_0 = - (A_1 v_{i1} + A_2 v_{i2} + A_3 v_{i3})$$

## Integrator



$$V_0(t) = -\frac{1}{RC} \int V_i(t) dt$$

$$RC = \text{Integrating time constant (s)}$$

No: \_\_\_\_\_

Date: \_\_\_/\_\_\_/\_\_\_

Q<sub>3</sub>

Sketch the o/p waveform from  $t=0$  onwards for,

(a)  $V_i = 1V$

(b)  $V_i = -1V$

(c) for a square wave i/p

Assume  $R = 1M\Omega$   
 $F = 1\mu F$  }  $RC = 1\mu s$

